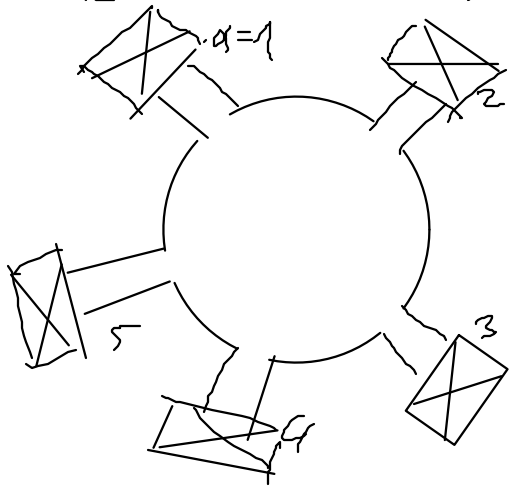


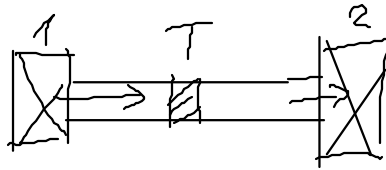
Landauer - Bittier

$$\frac{\hbar}{2e} I_\alpha = [N_\alpha - R_\alpha] \mu_\alpha - \sum_{\beta \neq \alpha} T_{\alpha\beta} \mu_\beta$$



$$N_\alpha = \sum_{\beta \neq \alpha} T_{\beta\alpha} + R_\alpha$$

Einfache Beispiele



1) 2 Kontakt-System

$$I_1 = I_2 = I ; T_{12} = T_{21} = T ; R_1 = R_2 = R = N - T$$

$$\frac{\hbar}{2e} I = (N - R) \mu_1 - T \mu_2 \quad \text{I}$$

$$-\frac{\hbar}{2e} I = (N - R) \mu_2 - T \mu_1 \quad \text{II}$$

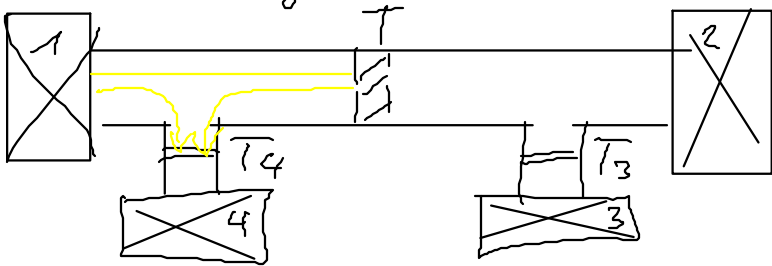
Differenz: $2 \frac{\hbar}{2e} I = (N - R) (\mu_1 - \mu_2) + T (\mu_1 - \mu_2)$

$$= (N - R + T) (\mu_1 - \mu_2) = 2T (\mu_1 - \mu_2)$$

$$\Rightarrow I = \frac{2e}{\hbar} (\mu_1 - \mu_2) = G_{2p} \frac{\mu_1 - \mu_2}{e} \quad \text{mit } G_{2p} = \frac{2e^2}{\hbar}$$

Summe: $0 = (N - R) (\mu_1 + \mu_2) - T (\mu_1 + \mu_2) = 0 \checkmark$

2) 4-Kontakt-System mit 1 Kanal



Die Spannungskontakte
wollen schwach angelegt
sein: $T_3, T_4 \ll 1$

Zur Vereinfachung $T_3 = T_4 = 0, R_3 = R_4$

$$N_i = 1$$

$$\textcircled{1} \frac{\hbar}{2e} I = (1 - R_1) \mu_1 - T_{12} \mu_2 - T_{13} \mu_3 - T_{14} \mu_4$$

$$\textcircled{2} -\frac{\hbar}{2e} I = (1 - R_2) \mu_2 - T_{21} \mu_1 - T_{23} \mu_3 - T_{24} \mu_4$$

$$\textcircled{3} 0 = (1 - R_3) \mu_3 - T_{31} \mu_1 - T_{32} \mu_2 - T_{34} \mu_4$$

$$\textcircled{4} 0 = (1 - R_4) \mu_4 - T_{41} \mu_1 - T_{42} \mu_2 - T_{43} \mu_3$$

Abschätzung

$$T_{12} = T_{21} = T + O(\epsilon)$$

$$T_{13} = T \cdot \epsilon = T_{31} = T_{24} = T_{42}$$

$$T_{14} = \epsilon \cdot R \cdot T = T_{41} = T_{23} = T_{32}$$

$$T_{34} = \epsilon^2 T = T_{43} \ll 1 \text{ wird vernachlässigt}$$

$$\textcircled{3} - \textcircled{4}: 0 = (1 - R_3)(\mu_3 - \mu_4) - T_{31}(\mu_1 - \mu_2) - T_{32}(\mu_2 - \mu_1)$$

$$0 = (T_{13} + T_{23} + T_{43})(\mu_3 - \mu_2) - (T_{32} - T_{31})(\mu_2 - \mu_1)$$

$$\Rightarrow \mu_3 - \mu_4 = \frac{T_{32} - T_{31}}{T_{32} + T_{31}} (\mu_2 - \mu_1) = \frac{\epsilon(1+R) - \epsilon T}{\epsilon(1+R) + \epsilon T} (\mu_2 - \mu_1)$$

$$= \frac{2R}{2} (\mu_2 - \mu_1)$$

$$\mu_3 - \mu_4 = R(\mu_2 - \mu_1)$$

$$\textcircled{1} - \textcircled{2}: \frac{2R}{2\epsilon} I = (1 - R)(\mu_1 - \mu_2) - T_{12}(\mu_2 - \mu_1) + O(\epsilon)$$

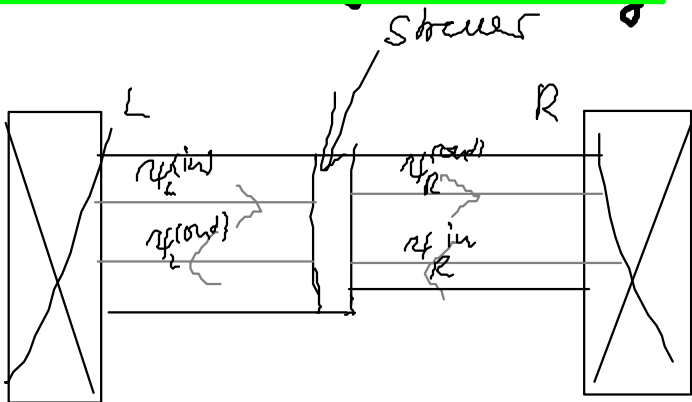
$$= (1 - R + T)(\mu_1 - \mu_2)$$

$$= 2T(\mu_1 - \mu_2)$$

$$\Rightarrow G_{2P} = \frac{eI}{\mu_1 - \mu_2} = \frac{e\epsilon^2}{R} T$$

$$G_{4P} = \frac{eI}{\mu_4 - \mu_3} = \frac{eI}{R(\mu_1 - \mu_2)} = \frac{2e\epsilon^2}{R} T$$

2.8 S-Matrix für Streuung



$$\psi_L^{\text{in}}(x, y) = \sum_{n=1}^{N_L} a_n^L \chi_n^L(y) e^{ik_n x}$$

$$\psi_L^{\text{out}}(x, y) = \sum_{n=1}^{N_L} b_n^L \chi_n^L(y) e^{-ik_n x}$$

$$\psi_R^{\text{in}}(x, y) = \sum_{m=1}^{N_R} a_m^R \chi_m^R(y) e^{ik_m x}$$

$$\psi_R^{\text{out}}(x, y) = \sum_{m=1}^{N_R} b_m^R \chi_m^R(y) e^{ik_m x}$$

• S ist eine quadratische $(N_L + N_R)^2$ Matrix

$b = S a$ a und b sind $(N_L + N_R)$ -Vektoren

oder " $\psi^{\text{out}} = S \psi^{\text{in}}$ "

• Das ganze kann auf mehrere Kontakte verallgemeinert werden

$$S \Leftrightarrow t_{\alpha\beta}, r_{\alpha\beta}; \tau_{\alpha\alpha}, r_{\alpha\alpha}$$

S-Matrix

$$\begin{pmatrix} b_1^L \\ b_2^L \\ \vdots \\ b_{N_L}^L \\ b_1^R \\ \vdots \\ b_{N_R}^R \end{pmatrix} = \begin{pmatrix} \Gamma_{11} & \Gamma_{12} & \dots & \Gamma_{1N_L} & | & t_{11}^L & \dots & t_{1N_R}^L \\ \vdots & \ddots & & \vdots & | & \vdots & & \vdots \\ \Gamma_{N_L 1} & \dots & \Gamma_{N_L N_L} & | & t_{N_L 1}^L & \dots & t_{N_L N_R}^L \\ \vdots & & \vdots & | & \vdots & & \vdots \\ t_{11}^R & \dots & t_{1N_L}^R & | & \Gamma_{11} & \dots & \Gamma_{1N_R} \\ \vdots & & \vdots & | & \vdots & & \vdots \\ t_{N_R 1}^R & \dots & t_{N_R N_L}^R & | & \Gamma_{N_R 1} & \dots & \Gamma_{N_R N_R} \end{pmatrix} \begin{pmatrix} a_1^L \\ a_2^L \\ \vdots \\ a_{N_L}^L \\ a_1^R \\ \vdots \\ a_{N_R}^R \end{pmatrix}$$

Eigenschaften

a) S ist unitär $SS^\dagger = S^\dagger S = \mathbb{1}$

$$\Rightarrow \langle \psi^{in} | \psi^{in} \rangle = \int_{-\frac{L}{2}}^{\frac{L}{2}} dx dy \sum_{n=1}^{N_L+N_R} \sum_{n'=1}^{N_L+N_R} a_n^* a_{n'} \chi_n^*(y) \chi_{n'}(y) e^{i k_n x \pm i \epsilon_n t} e^{\pm i k_{n'} x \pm i \epsilon_{n'} t}$$

$$= L \sum_n |a_n|^2$$

$$\langle \psi^{out} | \psi^{out} \rangle = L \sum_n |b_n|^2 = L \sum_{n, n'} S_{nn'}^* a_{n'}^* S_{nn} a_n$$

$$= L \sum_{n, n'} S_{n'n}^+ S_{nn'}^* a_n^* a_{n'}$$

mit $\psi^{in} = \begin{pmatrix} \psi_R^{in} \\ \psi_L^{in} \end{pmatrix}$

$$= L \sum_{n, n'} \underbrace{(S^\dagger S)_{n'n}}_{\delta_{n'n}} a_n^* a_{n'} = L \sum_n |a_n|^2$$

$$= \langle \psi^{in} | \psi^{in} \rangle$$

• Aus Unitarität von S folgt Norm, aber wir können dies nicht umdrehen und damit beweisen, dass S unitär sein muss.

Behauptung: S ist unitär folgt aus Stromerhaltung

$$I_L^{in} = \text{Re} \left[\frac{e\hbar}{m} \int dx \psi_L^{in*} (-i\hbar \nabla_x) \psi_L^{in} \right]$$

$$= \frac{e\hbar}{m} \sum_{n=1}^{N_L} |a_n|^2 k_n$$

$$I_L = I_L^{in} - I_L^{out} = I_R = I_R^{out} - I_R^{in} \quad (\text{Stromerhaltung}) \quad \text{rechts = links}$$

$$\Rightarrow I_L^{in} - I_L^{out} + I_R^{in} = I_L^{out} + I_R^{out} = I^{out} \quad \text{in} = \text{out}$$

$$\Rightarrow \sum_{n=1}^{N_L+N_R} |a_n|^2 k_n = \sum_{n=1}^{N_L+N_R} |b_n|^2 k_n$$

$$\frac{w}{c\hbar} I^{\text{out}} = \sum_{m=1}^n |b_m|^2 k_m = \sum_{m=1}^n \sum_{n=1}^n S_{mn}^* a_n^* S_{mn} a_n k_m$$

$$= \sum_{n=1}^n \sum_m S_{nm}^+ S_{mn} k_m a_n^* a_n$$

• Führe neue Matrix ein: $\tilde{S}_{mn} = S_{mn} \sqrt{\frac{k_m}{k_n}}$

$$S_{mn} = \sqrt{\frac{k_n}{k_m}} \tilde{S}_{mn}$$

$$S_{mn}^+ = S_{nm}^* = S_{nm}^* \sqrt{\frac{k_n}{k_m}} = S_{nm}^+ \sqrt{\frac{k_n}{k_m}}$$

$$S_{mn}^+ = \tilde{S}_{mn}^+ \sqrt{\frac{k_m}{k_n}}$$

$$\frac{w}{c\hbar} I^{\text{out}} = \sum_{n=1}^n \sum_m S_{nm}^+ \sqrt{\frac{k_n}{k_m}} \tilde{S}_{mn}^+ \sqrt{\frac{k_m}{k_n}} k_m a_n^* a_n$$

$$= \sum_{n=1}^n \left(\sum_m S_{nm}^+ \tilde{S}_{mn}^+ \right) \sqrt{k_n k_n} a_n^* a_n$$

wenn $\tilde{S}^+ \tilde{S} = \mathbb{1}$ also \tilde{S} unitär

$$= \sum_n |a_n|^2 k_n \text{ Stromerhaltung} \quad \Downarrow$$

• Was bedeutet dies für S ?

$$\sum_m S_{nm}^+ S_{mn} = \sum_m \tilde{S}_{nm}^+ \sqrt{\frac{k_m}{k_n}} \sqrt{\frac{k_n}{k_m}} \tilde{S}_{mn}$$

$$= \sqrt{\frac{k_n}{k_n}} (\tilde{S}^+ \tilde{S})_{nn} = \mathbb{1}$$

b) Für Zeitumkehrinvariante Probleme $\Rightarrow S = S^T \Rightarrow t_{mn} = t_{nm}$

$$t_{q\beta, \alpha n} = t_{p\alpha, mn}$$

$$G_{\alpha\beta} = G_{\beta\alpha}$$

Im Magnetfeld keine Zeitumkehrinvarianz

$$S(\vec{B}) = S^T(-\vec{B}) \Rightarrow t_{q\beta, mn}(\vec{B}) = t_{p\alpha, mn}(-\vec{B})$$

$$\Rightarrow G_{\alpha\beta}(\vec{B}) \neq G_{\alpha\beta}(-\vec{B})$$